

Lecture #3: Sublinear Fourier Algorithms  
Sparse

①

②

- All of the previous algs we discussed were inefficient. Yes, they are sample efficient but not computationally; we take considerably

fewer samples than the full signal but then do lots of FFTs! All to return only  $k$  items! Very inefficient (unless your application is really only constrained by sampling time/power space/...).

- Let's look at architecture of CoSaMP to understand the bottlenecks and what we should change:

$$① r = y - \Phi x^{(t)} = \Phi(x - x^{(t)}) = \text{sample from residual signal}$$

or

sample from current approx " and subtract from measurements.

Don't compute  $\Phi x^{(t)}$  explicitly!

$$② w^{(t+1)} = \underset{\substack{z \in \mathbb{C}^n \\ \text{supp } z \subset \Lambda^{(t+1)}}}{\operatorname{argmin}} \|y - \Phi z\|_2$$

$$\mathcal{O}(k \lg n) = m \quad \begin{matrix} 3k \\ \boxed{z} \\ \Lambda^{(t+1)} \end{matrix} = \boxed{y} \quad \text{solve this over-determined system. It's small!}$$

OR - knowing  $\Lambda^{(t+1)} = 3k$ -term supp set, estimate the vals of the entries of  $z$  on that support set.

$$③ \text{Supp}_{2k} (\Phi^* r)$$

these are our two big bottlenecks

given samples (of residual), how to find/identify top  $2k$  entries WITHOUT computing  $\Phi^* r$

- This is actually revisionist history! SFT algos have been around for a while, just in slightly different form.

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1991 Kushilevitz & Mansour  
 1989/1993 Goldreich & Levin  
 (Crypto/codes)

(ML)  
 Hadamard transform  
 (Fourier analysis on Boolean cube)

1992 Kushilevitz (ML)  
 2002 GGIMS (streaming/sublinear alg)  
 2003 Akavia, Goldwasser,  
 Safra (crypto.)

Fourier analysis on  $\mathbb{Z}_n$   
 for various flavors of  $n$   
 $(n \text{ prime}, n=2^l, \dots)$

→ 2005 Gilbert, Muthukrishnan,  
 Strauss

Randomized algos  
 with constant prob. of  
 error (i.e., per signal)

$L_1/L_2$  results

$$\begin{aligned} \text{running time} \\ (+ \# \text{samples}) = O(k \lg(n)) \end{aligned}$$

2010 - onwards lots of work!

big flumy of algos, implementations,  
 and hardware.

Det'ic constructions, too.

See 2015 SP Mag survey, Iwen, Indyk, Schmidt.

With

most prominent one is HIKP2012 b = Hassanein, Indyk, Katabi, Price.

Let's outline basic components & techniques (that don't suffer from the bottlenecks).

① To get intuition, let's assume spectrum of  $x$  (on  $\mathbb{Z}_n$ ) consists of a

SINGLE non-zero freq.

- two samples are sufficient to get position and value/coeff.  
 of freq.

-  $\lg(n)$  samples suffice if tone + noise.

↳ i.e., build a filterbank

② If we have  $\dots \rightarrow$  from  $\mathbb{R}^n$  to  $\mathbb{C}^m$  then estimate of Fourier space

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- if each bin has only ONE freq from the signal in it, then  
the filter bank succeeds in isolating the significant freqs,  
and, if we can sample from such a filtered signal, we  
have reduced our problem to case ①.

→ sample complexity & run time is proportional to # bins.

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for close freqs.,  
we permute the  
spectrum randomly  
and then apply  
filterbank.

① Single freq. recovery : suppose  $x = \sum_{w \in \mathbb{Z}_n} \alpha_w e^{j\omega w j/n}$  with coeff  $\alpha_w \in \mathbb{C}$ .

$$x_j = \alpha_w \frac{1}{\sqrt{n}} e^{-j\omega w j/n} \quad j \in \mathbb{Z}_n$$

a. calculate  $\omega$  by choosing  $j, j+1 \in \mathbb{Z}_n$  and calculating

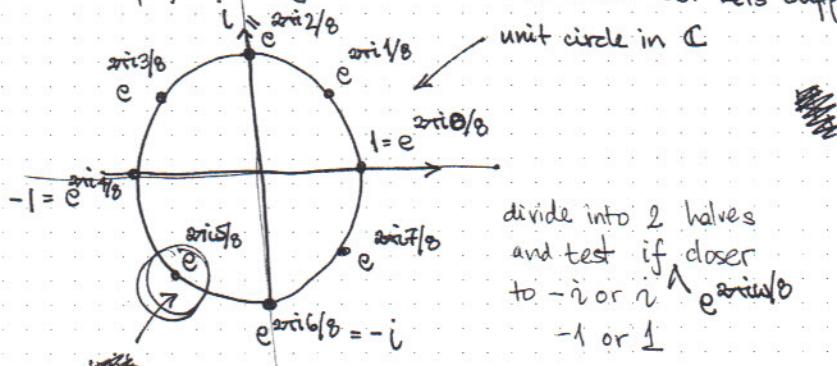
$$\text{phase} \left( \frac{x_{j+1}}{x_j} = \frac{\alpha_w \sqrt{n}}{\alpha_w \sqrt{n}} e^{-j\omega w(j+1)/n} = e^{-j\omega w/n} \right)$$

$$= \text{phase} \left( \cos \left( \frac{-2\pi \omega}{n} \right) + i \sin \left( \frac{-2\pi \omega}{n} \right) \right).$$

b. once you know  $\omega$ , calculate  $\alpha_w$  by computing

$$x_j \sqrt{n} e^{-j\omega w j/n} = \alpha_w \cdot 1.$$

this doesn't work very well if  $x$  is noisy. Have to  
perform binary search to zoom in on  $\omega$ . Let's suppose  $n=8$ .



$$(i) |x_j| \cdot |i - e^{\frac{2\pi i w}{8}}| = |i \cdot x_j - x_{j+1}| < |i \cdot x_j + x_{j+1}| = |x_j| \cdot |i + e^{\frac{2\pi i w}{8}}|$$

$$(ii) |x_j| \cdot |1 - e^{\frac{2\pi i w}{8}}| = |x_j - x_{j+1}| < |x_j + x_{j+1}| = |x_j| \cdot |1 + e^{\frac{2\pi i w}{8}}|$$

if  $|1 - e^{\frac{2\pi i w}{8}}| < |1 + e^{\frac{2\pi i w}{8}}|$ , (i.e.,  $e^{\frac{2\pi i w}{8}}$  is closer to  $+1$ )  
then we'll see

$$|x_j - x_{j+1}| < |x_j + x_{j+1}| \quad \left[ \text{and similarly for } \pm i \right]$$

If  $e^{\frac{2\pi i w}{8}}$  is closer to  $+1$ , then  $w \in \{7, 0, 1\}$  and  $w \notin \{3, 4, 5\}$

[similarly for  $\pm i$  and  $\{5, 6, 7\}$  and  $\{1, 2, 3\}$ .]

let's suppose we learn  $w \in \{4, 5, 6\}$ , then we can simplify our problem

$$\text{define } x'_j = e^{-2\pi i \cdot 4 \cdot (2j/8)} x_{2j}$$

$$= e^{-2\pi i \cdot 4 \cdot (2j/8)} \frac{\alpha_w}{\sqrt{n}} e^{2j\pi i w/8}$$

$$= \frac{\alpha_w}{\sqrt{n}} \cdot e^{2\pi i (w-4)/4}$$

for  $j \in \mathbb{Z}_{n/2} = \mathbb{Z}_4$ .

domain is half as big.

$\Rightarrow$  halved our problem.

- we shifted/rotated domain into first quadrant by multiplying by  $e^{-2\pi i j/8}$

- then discarded odd entries.

and iterate...

this is in the Gilbert-Strauss-Tropp 2009 survey, albeit expressed diff'ly.

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Let's do an example. Suppose  $n = 2 \cdot 5 \cdot 7 = 70$   
and suppose

$$x_j = \alpha_w \cdot \frac{1}{\sqrt{n}} \cdot e^{\frac{2\pi i j w}{n}}$$

$\left[ \begin{array}{l} x \text{ is 1-tone with freq. } w \in \mathbb{Z}_n \\ \text{for } j \in \mathbb{Z}_n \end{array} \right]$

let's form a short vector  $a \in \mathbb{C}^2$  by sampling  $x$  at  $j=0, n/2$ .

$$a_0 = x_0 = \alpha_w \cdot \frac{1}{\sqrt{n}}$$

$$a_1 = x_{n/2} = \alpha_w \cdot \frac{1}{\sqrt{n}} e^{\frac{2\pi i w \cdot n/2}{n}} = \alpha_w \cdot \frac{1}{\sqrt{n}} (-1)^w$$

and compute the Fourier transform of  $a$  ( $\hat{a} \in \mathbb{C}^2$ ):

$$\hat{a}_0 = \alpha_w \cdot \frac{1}{\sqrt{n}} \left( \frac{1 + (-1)^w}{\sqrt{2}} \right)$$

$$\hat{a}_1 = \alpha_w \cdot \frac{1}{\sqrt{n}} \left( \frac{1 + (-1)^{w+1}}{\sqrt{2}} \right)$$

w is an integer and  
it's either EVEN or ODD  
⇒ only ONE of  $\hat{a}_0$  or  $\hat{a}_1$   
will be non-zero  
if  $\hat{a}_0 \neq 0$ , then  $w \equiv 0 \pmod{2}$   
if  $\hat{a}_1 \neq 0$ , then  $w \equiv 1 \pmod{2}$

Can learn  $w \pmod{5}$  and  $\pmod{7}$  by sampling  $x$  at intervals  $n/5, n/7$

~~and then computing shorter FFTs.~~ [aliased FFTs  
can do in parallel!]

Then, from moduli, can reconstruct  $w$ .

### Thm (CRT)

Any integer  $x$  is uniquely specified mod  $n$  by its remainders  
mod  $m$  rel. prime ints  $p_1, p_2, \dots, p_m$  as long as  $\prod_{i=1}^m p_i \geq n$ .

→ Q7: How many samples? How fast?

$$2 + 5 + 7 = 14$$

↓

FFT

the algorithm is a  
small linear system  
size = # prime factors.  
 $\sim O_n(n)$

## 2. Filtering to Isolate Freqs.

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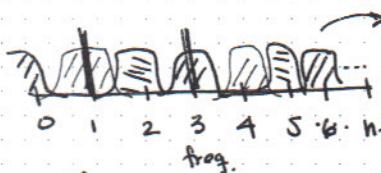
It's important to view our CRT argument as a sampling from a filtered version of our signal in order to figure out how to handle signals with more than 1 freq.



0 1 2 3 4 ... n-1  
freq.

→ the sampling at 0,  $n/2$   
"filters" the signal into 2 freq. "bins"—  
those that are EVEN & ODD.

Similarly for the other factors  
bins for different conj. classes mod 5, 7.



0 1 2 3 4 5 6 ... n-1  
freq.  
If signal has 2 freqs, then the freqs. can't be equal mod 3  
(CRT says they can't collide if you take enough primes)

ex:

3 bins for 0, 1, 2 mod 3.

2 freqs in spectrum.

More generally: need ~~at least~~  $k^3$   
bins to isolate the  $k$  freqs.  
⇒ ~~at least~~ small # of short FFTs



→ lots of research on good  
near-perfect bandpass filters.  
e.g. sinc \* Gaussian

[need a short # coeffs on time...]

diff. sublinear algos. have used  
diff. filters: Gaussians, Dolph-Chirpysch.,  
indicators, spike trains, ...

give enough info. to get  
all  $k$  freqs.

Q7: How do we handle freqs. that are close together?

These bins are det'ic!

but sample eff'ly

Need:

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- ① permute spectrum efficiently  $\rightarrow$  use pairwise indep. random permutations
- ② sample on time side from the permuted spectrum  $\leftarrow$  in conjunct w/ use 2 invariance props. of the FT.

a) dilation

$$a_j = x_{cj} \longleftrightarrow \hat{a}_j = \hat{x}_{c^{-1}j} \text{ assuming } c^{-1} \text{ exists mod } n$$

b) translation

$$a_j = e^{\frac{2\pi i b j}{n}} \longleftrightarrow \hat{a}_j = \hat{x}_{j-b}$$

$\Rightarrow$  if we choose  $b, c$  indep, unif'ly at random ( $c \in \mathbb{Z}_n^*$ ), then

$$a_j = e^{\frac{2\pi i b j}{n}} x_{cj}$$

has a permuted spectrum  $\hat{a}_j = \hat{x}_{c^{-1}j-b}$ ; i.e.,

if  $\hat{x}_w = \hat{x}_w$ , then  $\hat{a}_{(wc+b) \bmod n} = x_w$

we permute  
the spectrum

You check: ① if  $b, c$  indep, unif. random ( $c \in \mathbb{Z}_n^*$ ), then

$w \mapsto wc + b$  is a random permutat'

② it's actually a pairwise indep. rand. permutat'

$$\begin{aligned} cw_1 + b &= \xi_1 \\ cw_2 + b &= \xi_2 \end{aligned} \quad \left. \begin{array}{l} \text{what's the prob.} \\ \text{of collision?} \end{array} \right\}$$

Put all these pieces together in an iterative procedure:

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while not done {

  IDENTIFY freqs. with big Fcoeffs

  ESTIMATE Fcoeffs of those identified freqs.

  SUBTRACT (samples of) current approx. from  
    (samples of) orig. signal. (i.e., remeasure).

} [GST 2008 has full pseudo code  
that is easy to implement!]  
(note caveats!)

There are some implementations:

  AAPP (on sourceforge)

  [ HIKP1,2 also have pseudocode ]

  SFFT v1,2,3 } (on MIT webpage)

  ETH

seeIEEE SP Mag 2008, GST  
for a nicer layout  
+ diagrams!SFT( $x, k$ )

Input:  $x$  has length  $n = 2^l$  (need sampling access or gen. samples FIRST and input those)  
 $k$  # of desired freqs.

Output:  $\Delta = \{(w, \alpha_w)\}$  = list containing  $O(k)$  (freq, coeff) pairs

$K \leftarrow 8k$  ] Initialize:  
 $\Delta \leftarrow \emptyset$

for  $j=1$  to  $S$  {

$\Omega \leftarrow \text{Identification}(x, \Delta, K)$

$c \leftarrow \text{Estimation}(x, \Delta, \Omega)$

for each  $w \in \Omega$  {

if  $(w, \alpha_w) \in \Delta$  ~~then~~, then replace with  $(w, \alpha_w + c_w)$  in  $\Delta$

} else add new pair  $(w, c_w)$  to  $\Delta$

Retain  $K$  pairs in  $\Delta$  with largest (abs. val.) coeff

}

Return  $\Delta$  or prune to top  $k$  pairs.

Sample-Residual( $x, \Delta, t, \sigma, K$ )

for  $k=1$  to  $K$  {

$u_k \leftarrow x(t + \sigma(k-1) \bmod n)$  ] sample  
 arith. prog. from signal

$v_k \leftarrow \sum_{(w, \alpha_w) \in \Delta} (d_w e^{2\pi i w t / n}) e^{2\pi i w \sigma / n (k-1)}$  ] non-uniform FFT

}

Return  $u - v \rightarrow$  residual

## Identification ( $x, \Lambda, K$ )

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$\text{reps} \leftarrow s$   
 $w_k \leftarrow 0 \text{ for } k=1,2,\dots,K$

initialize

Draw  $\sigma \sim \text{Unif}\{1, 3, 5, \dots, n-1\}$   $\rightarrow$  random odd dilate factor  
(needs to be odd so it's invert.  
 $\text{mod } 2^k$ ).

for  $b = 0$  to  $\lg(n/2)$  { loop from LSB to MSB

$\text{vote}_k \leftarrow 0 \text{ for } k=1,2,\dots,K$

for  $j=1$  to  $\text{reps}$  {

Draw  $t \sim \text{Unif}\{0, 1, \dots, n-1\}$   $\rightarrow$  random sample pt.

$u \leftarrow \text{sample-shattering}(x, \Lambda, t, \sigma, K)$

$v \leftarrow \text{Sample-Shattering}(x, \Lambda, t + n/2^{b+1}, \sigma, K)$

for  $k=1$  to  $K$  {

$$E_0 \leftarrow u_k + e^{-\pi i w_k / 2^b} v_k$$

$$E_1 \leftarrow u_k - e^{-\pi i w_k / 2^b} v_k$$

if  $|E_1| \geq |E_0|$ , then  $\text{vote}_k \leftarrow \text{vote}_k + 1$

vote when bit = 1

samples for correlated testing  $b^{th}$  bit

apply bit-test to demod. sig.

for  $k=1$  to  $K$  {

if  $\text{vote}_k > \text{reps}/2$  then  $w_k \leftarrow w_k + 2^b$

majority vote for bit value

}  
Return  $\text{unique}(w_1, w_2, \dots, w_K)$   $\rightarrow$  remove duplicate freqs.

Sample-Shattering ( $x, \Delta, t, \sigma, K$ )

$z \leftarrow \text{Sample-residual}(x, \Delta, t, \sigma, K)$

$z \leftarrow \text{FFT}(z)$

return ( $z$ )

Estimation ( $\kappa, \Delta, \Omega$ )

reps  $\leftarrow 5$

for  $j=1$  to reps {

Draw  $\sigma \sim \text{Unif.}\{1, 3, 5, \dots, n-1\}$

$t \sim \text{Unif.}\{0, 1, 2, \dots, n-1\}$

$n \leftarrow \text{Sample-residual}(x, \Delta, t, \sigma, K)$

for  $l=1$  to  $|\Omega|$  {

$$c_l(j) = \sum_{k=1}^K u_{lk} e^{\frac{2\pi i (\omega_l \sigma/n)(k-1)}} \quad \begin{array}{l} \text{in parallel,} \\ \text{non-unif FFT} \end{array}$$

$$c_l(j) \leftarrow \left(\frac{n}{K}\right) e^{-2\pi i \omega_l t/n} c_l(j) \quad \begin{array}{l} \text{demodulate and} \\ \text{scale the est.} \end{array}$$

}

$$\dots c_\Omega \leftarrow \text{Median}\{c_l(j) \mid j=1, \dots, \text{reps}\} \text{ for } l=1, 2, \dots, |\Omega|$$

return  $c_1, c_2, \dots, c_{|\Omega|}$

helps with robustness.